plex that it has never been successfully used for photomultiplier dynodes.

The "bialkali" (K_2 CsSb) photocathode, first described by this author, ² has apparently never been examined for its secondary emission characteristics although it is widely used as a blue-sensitive cathode of high efficiency. Experiments to measure the secondary emission gain from K_2 CsSb have now been performed, and the results indicate that this material has much higher gain factors than Cs_3 Sb. Figure 1 shows curves of gain as a function of primary energy for four materials, viz., the conventional Cs_3 Sb and S-20 [(Cs)Na₂KSb] cathodes and the new Cs_2 Sb material both before and after superficial oxidation. The two other conventionally used dynode materials, MgO and BeO (used in the form of oxidized Ag-Mg and Cu-Be alloys, respectively), have even lower gain factors than Cs_3 Sb.

In a qualitative way, the higher gain of K_2CsSb , as well as that of the S-20 cathode material, as compared with that of Cs_3Sb is consistent with the photoemissive characteristics of these materials since the peak quantum efficiency of both "multialkali" materials is two or three times greater than that of Cs_3Sb .

The delay in the discovery of the high gain of K₂CsSb has reduced the impact that such an improvement over

commercially used dynode materials would have had as recently as three years ago. Simon and Williams, 3 using the new concept of "negative effective electron affinity", demonstrated that heavily p-type doped GaP, after a surface activation with Cs, has extremely high secondary emission gains at high primary energies. Figure 2 shows a comparison of a typical GaP(Cs) dynode with a K₂CsSb dynode. It is apparent that the two materials are quite similar at low primary energies but that the gain of GaP continues to rise almost linearly with primary energy up to much higher values. This, as explained in Ref. 3, is due to the much greater escape depth from materials with negative electron affinity. At present, the GaP(Cs) dynode is used in photomultipliers^{4,5} at primary electron energies of approximately 600 V. From Fig. 2 it is apparent that at this voltage the gain of superficially oxidized K2CsSb is almost equally high.

On the Signal-to-Noise Ratio for Optical Heterodyne Detection*

M.C. Teich and R.Y. Yen

Department of Electrical Engineering and Computer Science, Columbia University, New York, New York 10027 (Received 11 October 1971; in final form 21 January 1972)

The conventional result for the signal-to-noise ratio in an optical heterodyne detector is shown to apply to the mixing of a stable local oscillator with any radiation field which gives rise to a stationary narrow-band i.f. signal. The result is explicitly demonstrated for radiation scattered from a diffuse target, which may be represented as a narrow-band Gaussian random process.

It has been shown elsewhere¹⁻³ that the ac response for an optical or ir heterodyne detector may be written as

$$r_{\text{i.f.}} \propto \mathcal{L}\left\{ (E_{\text{LO}}\cos\omega_{\text{LO}}t)(E_{\text{S}}(t)\cos[\omega_{\text{S}}t + \delta(t)]) \right\}$$
 (1a)

$$=\beta E_{LO}E_{S}(t)\cos[\omega_{i,f},t+\delta(t)], \qquad (1b)$$

provided that the heterodyne signal is a narrow-band process. Here, $r_{i.f.}$ represents the photodetector response at the intermediate frequency (i.f.), E_{LO} and E_{S} represent the magnitude of the electric field for the local oscillator (LO) and signal (S) beams, respectively, ω is the angular frequency of the particular radiation beam, δ is a phase angle, and β is a proportionality constant containing the detector quantum efficiency. The operator £ stands for "the low frequency part of", and the quantity $\omega_{\text{i.f.}}$ is defined as $|\omega_{\text{S}} - \omega_{\text{LO}}|$. If the LO is a well-stabilized single-mode laser operated well above threshold, $E_{\rm LO}$ and $\omega_{\rm LO}$ are strictly constant, and the quantities $E_S(t)$ and $\delta(t)$ reflect the amplitude and phase fluctuations of the signal beam. For simplicity, spatial coherence has been assumed over the detector aperture, and the addition of a constant phase factor has been omitted.

To obtain the power signal-to-noise ratio $(S/N)_{\rm power}$ for a narrow-band stationary process such as described by Eq. (1b), we first square and then time average this equation to obtain

$$\langle r_{i,f.}^2 \rangle = \frac{1}{2} \beta^2 E_{LO}^2 \left\{ \langle E_S^2(t) \rangle + \langle E_S^2(t) \cos[2\omega_{i,f.}t + 2\delta(t)] \rangle \right\}.$$
 (2)

Because $E_s(t)$ and $\delta(t)$ are independent random variables, the second term above may be written as

$$\langle E_S^2 \cos[2\omega_{i,f}t + 2\delta] \rangle = \langle E_S^2 \rangle \langle \cos[2\omega_{i,f}t + 2\delta] \rangle = 0.$$
 (3)

since $\langle\cos[2\omega_{i.f.},t+2\delta]\rangle=0$ for the phase uniformly distributed over $(0,\ 2\pi)$. Furthermore, for a strong LO, which we assume here, the dc part of the detector response may be written as $r_{\rm dc}\!\simeq\!\frac{1}{2}\beta E_{\rm LO}^2$. Thus, Eq. (2) becomes

$$\langle r_{1.f.}^2 \rangle = \frac{1}{2} \beta^2 E_{LO}^2 \langle E_S^2 \rangle = 2 \langle E_S^2 \rangle r_{dc}^2 / \langle E_{LO}^2 \rangle, \tag{4}$$

which is also given by

$$\langle r_{i.t.}^2 \rangle = 2P_S r_{dc}^2 / P_{LO}$$
 (5)

where the quantities $P_S \propto \langle E_S^2 \rangle$ and $P_{\text{LO}} \propto \langle E_{\text{LO}}^2 \rangle$ represent

J. Appl. Phys., Vol. 43, No. 5, May 1972

¹A.A. Mostovskii, G.B. Vorobeva, and G.B. Struchinskii, Sov. Phys. Solid State 5, 2436 (1964).

²A. H. Sommer, Appl. Phys. Letters 3, 62 (1963).

³R.E. Simon and B.F. Williams, IEEE Trans. Nucl. Sci. NS-15, 167 (1968).

⁴R.E. Simon, A.H. Sommer, J.J. Tietjen, and B.F. Williams, Appl. Phys. Letters 13, 355 (1968).

⁵G. A. Morton, H.M. Smith, and H.R. Krall, Appl. Phys. Letters **13**, 356 (1968).

the radiation powers of the signal and LO beams, respectively.

If we now consider the noise response in the detector to arise from a simple shot process, the mean-square noise response $\langle r_n^2\rangle$ is given by the well-known shot noise formula 1,4

$$\langle r_{\rm r}^2 \rangle = 2er_{\rm do}\Delta f,\tag{6}$$

where Δf is the bandwidth of the receiver. Thus, the $(S/N)_{\mathrm{power}}$ may be written as

$$\left(\frac{S}{N}\right)_{\text{power}} = \frac{\langle \gamma_{1,\text{f.}}^2 \rangle}{\langle \gamma_{2}^2 \rangle} = \frac{P_S}{e\Delta f} \frac{\gamma_{\text{dc}}}{P_{\text{LO}}}.$$
(7)

However, since $r_{\rm dc}$ arises from the comparitively large LO, it is related to the LO beam power $P_{\rm LO}$ by the detector quantum efficiency η :

$$r_{\rm dc} = (\eta e/h\nu) P_{\rm LO}. \tag{8}$$

Therefore, the signal-to-noise ratio becomes

$$(S/N)_{\text{power}} = \eta P_S / h \nu \Delta f, \tag{9}$$

which is the usual result obtained for the mixing of two sinusoidal signals. Thus, the validity of Eq. (9) has been extended to the mixing of a stable (sinusoidal) LO with a narrow-band stationary signal. Actually, the optimum S/N for a detector other than a photoemitter (e.g., a photoconductor) may differ by a constant from this expression. ^{1,4} Quite recently, van der Ziel⁵ has reported an expression for $(S/N)_{power}$ which takes into account the effect of mixing of the LO with the fluctuations in the incoming radiation, and differs from Eq. (9) by the factor $(1+2\eta)^{-1}$. These constant factors apply equally well to sinusoidal mixing and to mixing with a narrow-band signal, however.

One particular narrow-band signal of interest arises

from the heterodyning (or homodyning) of radiation scattered from a diffuse target with a stable LO, in which case Eq. (1b) would represent a narrow-band Gaussian process centered at $\omega_{i.t.}$.^{1,2} It is not difficult to explicitly show that Eq. (4), and therefore Eq. (9), are valid for this case. Letting $y=r_{i.t.}^2$, the probability density function p(y) is then first-order chi-square and is given by⁶

$$p(y) = [(2\pi y)^{1/2} \beta E_{10} \sigma_{S}]^{-1} \exp(-y/2\beta^{2} E_{10}^{2} \sigma_{S}^{2}). \tag{10}$$

Here σ_S is the variance of the narrow-band Gaussian signal. The mean $\langle y \rangle$ is readily evaluated to be

$$\langle y \rangle = \langle r_{i,f}^2 \rangle = \beta^2 E_{i,f}^2 \sigma_s^2. \tag{11}$$

Since the square of the signal radiation envelope $E_S^2(t)$ is exponentially distributed with a mean $\langle E_S^2 \rangle = 2\sigma_S^2$, we have

$$\langle r_{i,f_s}^2 \rangle = \frac{1}{2} \beta^2 E_{LO}^2 \langle E_s^2 \rangle,$$
 (12)

in agreement with Eq. (4).

^{*}Work supported by the National Science Foundation.

¹M.C. Teich, in *Semiconductors and Semimetals*, edited by R.K. Willardson and A.C. Beer (Academic, New York, 1970), Vol. 5, pp. 361-407.

²M.C. Teich, Proc. IEEE 57, 786 (1969).

³M.C. Teich, Appl. Phys. Letters 14, 201 (1969); M.C. Teich, in *Proceedings of the Third Photoconductivity Conference*, edited by E.M. Pell (Pergamon, New York, 1971), pp. 1-5.

⁴M.C. Teich, Proc. IEEE 56, 37 (1968).

⁵A. van der Ziel, Proc. IEEE 58, 1178 (1970).

⁶W.B. Davenport, Jr. and W.L. Root, *An Introduction to the Theory of Random Signals and Noise* (McGraw-Hill, New York, 1958), p. 253.